EEE321 Electromagnetic Fields and Waves

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(11th Week)

Outline

- Electromagnetic Spectrum
- Uniform Plane Electromagnetic Waves in General Medium
- Uniform Plane Waves in Free Space
- Uniform Plane Waves in Nononductor
- Uniform Plane Waves in Conductor
- Poynting Theorem

Electromagnetic Spectrum

- Maxwell'e equations impose no limit on the frequency of electromagnetic waves
- Known spectrum exteds continuously from long radio waves to the very high-energy gamma rays of cosmic radiation.
- At the beginning the frequencies were the order of 100 hertz and the wavelength about 3 megameters
- Recenly the frequencies are of the order of 10²⁴ hertz and the wavelength less then 1 fentometer
- Many experiments demonstrate that all wave are transverse and they travel at the speed of light in free space
- we use H instead of B in dealing with electromagnetic waves.
 - ExH is power density and E/H is impedance

THE ELECTROMAGNETIC SPECTRUM



Gigahertz (GHz) 10-9 Terahertz (THz) 10-12 Petahertz (PHz) 10-15 Exahertz (EHz) 10-18 Zettahertz (ZHz) 10-21 Vottahertz (YHz) 10-24



Uniform Plane Electromagnetic Waves in a General Medium

- A wave front is a surface of uniform phase
- The wave front of a planar wave is planar
- A wave is uniform if a wave front is a surface of uniform phase and uniform amplitute
- In general medium that is homogeneous isotropic, linear and stationary (HILS) has parameters ε_r , μ_r and σ
- We assume a sinusidal wave travelling in the direction of the z-axis
- The wave is lineary polarized. So **E** and **H** are of the form $E = E_m \exp j(\omega t kz), \quad H = H_m \exp j(\omega t kz)^{\ddagger} \quad (28-1)$
- Where Em and H_m are vectors that are independent of the time and of the coordinates
- If there is no attenuation, the wave numver is real

$$k = \frac{\omega}{v} = \frac{2\pi}{\lambda} = \frac{1}{\lambda},$$
(28-2)

Relative Orientation of E, H, and k

- For this perticular field $\frac{\partial}{\partial t} = j\omega, \quad \nabla = \frac{\partial}{\partial x}\hat{x} + \frac{\partial}{\partial y}\hat{y} + \frac{\partial}{\partial z}\hat{z} = -jk\hat{z}. \quad (28-3)$
- We set $\rho_f = 0$. We also set $J_f = \sigma E_f$ (28-4)
- On the assumption that vxB is negligible compared to E, where v is the velocity of a conduction electron
- The Maxwel's equations then become $-jk\hat{z} \cdot E = 0$, $-jk\hat{z} \times E = -j\omega\mu H$, (28-5) $-jk\hat{z} \cdot H = 0$, $-jk\hat{z} \times H = \sigma E + j\omega \epsilon E$ (28-6)
- And then to $\hat{z} \cdot E = 0$, $E = -\frac{k}{\omega \epsilon + j\sigma} \hat{z} \times H$, (28-7) $\hat{z} \cdot H = 0$, $H = \frac{k}{\omega \mu} \hat{z} \times E$. (28-8)

Characteristic Impedance Z of a Medium

 The ratio E/H is the characteristic impedance Z of a medium of propogation:

$$Z = \frac{E}{H} = \frac{k}{\omega \epsilon - j\sigma} = \frac{\omega \mu}{k}.$$
 (28-9)

Wave Number k

 The value of k² follows from the equation given for impedance definition:

$$k^{2} = \omega^{2} \epsilon \mu - j \omega \sigma \mu = \omega^{2} \epsilon \mu \left(1 - j \frac{\sigma}{\omega \epsilon}\right), \qquad (28-10)$$
$$= \omega^{2} \epsilon_{0} \mu_{0} \epsilon_{r} \mu_{r} \left(1 - j \frac{\sigma}{\omega \epsilon}\right). \qquad (28-11)$$

• The σ terms account for Joule losses and attenuation

Wave Equations

We found nonhomogenous wave equation for E and B as:

$$\nabla^{2} \boldsymbol{E} - \boldsymbol{\epsilon}_{0} \boldsymbol{\mu}_{0} \frac{\partial^{2} \boldsymbol{E}}{\partial t^{2}} = \frac{\nabla \rho}{\boldsymbol{\epsilon}_{0}} + \boldsymbol{\mu}_{0} \frac{\partial \boldsymbol{J}}{\partial t}, \qquad (28-12)$$

$$\nabla^{2} \boldsymbol{B} - \boldsymbol{\epsilon}_{0} \boldsymbol{\mu}_{0} \frac{\partial^{2} \boldsymbol{B}}{\partial t^{2}} = \boldsymbol{\mu}_{0} \nabla \times \boldsymbol{J}. \qquad (28-13)$$

- If we set $\rho_f = 0$ and ϵ , μ , σ are parameters of the medium
 - $\nabla^2 E \epsilon \mu \frac{\partial^2 E}{\partial t^2} = \mu \frac{\partial J_f}{\partial t} = \mu \sigma \frac{\partial E}{\partial t},$ (28-14) $\boldsymbol{\nabla}^2 \boldsymbol{B} - \boldsymbol{\epsilon} \boldsymbol{\mu} \frac{\partial^2 \boldsymbol{B}}{\partial t^2} = -\boldsymbol{\mu} \boldsymbol{\sigma} \boldsymbol{\nabla} \times \boldsymbol{E} = \boldsymbol{\mu} \boldsymbol{\sigma} \frac{\partial \boldsymbol{B}}{\partial t}.$ (28-15) $\nabla^2 E - \epsilon \mu \frac{\partial^2 E}{\partial t^2} - \mu \sigma \frac{\partial E}{\partial t} = 0,$ (28-16) $\nabla^2 B - \epsilon \mu \frac{\partial^2 B}{\partial t^2} - \mu \sigma \frac{\partial B}{\partial t} = 0,$ (28-17) Or to use **H** instead of **B** $\nabla^2 H - \epsilon \mu \frac{\partial^2 H}{\partial t^2} - \mu \sigma \frac{\partial H}{\partial t} = 0.$ (28-18)Then wave equation for **E** (**H** is also similar) is given by: (28-19) $(-k^2 + \omega^2 \epsilon \mu - j \omega \sigma \mu) E = 0,$

Uniform Plane Electromagnetic Waves in Free Space

• In free space $\epsilon_r = 1$, $\mu_r = 1$, $\sigma = 0$. There is no attenuation and the wave number is;

 $k = \frac{1}{\lambda_0}$ (28-20) = $\omega (\epsilon_0 \mu_0)^{1/2}$. (28-21) The speed of ligth is;

 $c = \frac{\omega}{k} = \frac{1}{(\epsilon_0 \mu_0)^{1/2}} = 2.99792458 \times 10^8 \text{ meters/second.}$ (28-22)

The characteristic impedance of the vacuum is

$$Z_0 = \frac{E}{H} = \frac{k}{\omega\epsilon_0} = \frac{\omega\mu_0}{k} = \frac{1}{\epsilon_0 c} = \mu_0 c = \left(\frac{\mu_0}{\epsilon_0}\right)^{1/2} = 3.767303 \times 10^2 \approx 377 \text{ ohms.}$$
(28-25)

• Thus, since $B = \mu_0 H$ in free space, $\frac{E}{B} = \frac{1}{(\epsilon_0 \mu_0)^{1/2}} = c$, or E = Bc. (28-26)

- The electric and magnetic energy densities are equal: $\frac{\epsilon_0 E^2/2}{\mu_0 H^2/2} = \frac{\epsilon_0}{\mu_0} \left(\frac{\mu_0}{\epsilon_0}\right) = 1.$ (28-27)
- At any instand, the total energy fluctuates with z as in figure and its time averaged value at any point

$$E = E_m \cos(\omega t - kz), \qquad H = H_m \cos(\omega t - kz).$$

- Abondoning the phasor notation $\mathscr{C}' = \frac{\epsilon_0 E_{\text{rms}}^2}{2} + \frac{\mu_0 H_{\text{rms}}^2}{2} = \epsilon_0 E_{\text{rms}}^2 = \mu_0 H_{\text{rms}}^2.$ (28-28)
- The magnetute of the Poynting vector is $|\mathcal{S}| = |\mathbf{E} \times \mathbf{H}| = E_m H_m \cos^2(\omega t - kz).$ (28-30)
- Time aveared Poynting vector is $\mathscr{G}_{av} = \frac{1}{2} \operatorname{Re} (E \times H^*)$ (28-31)



- For a uniform plane wave in free space $\mathscr{G}_{av} = \frac{1}{2} \operatorname{Re} (EH^*) \hat{z}$ (28-32) $= \frac{1}{2} c \epsilon_0 |E_m|^2 \hat{z} = c \epsilon_0 E_{rms}^2 \hat{z} = \frac{E_{rms}^2}{Z_0} \hat{z}$ (28-33) $\approx \frac{E_{rms}^2}{277} \hat{z}$ watts/meter². (28-34)
- The time-aveaged Poyinting vector is equal to total energy density multiplied by the speed of ligth c

Uniform Plane Electromagnetic Waves in Nonconductor

• The stuation is the same as in free space, with ε and μ replacing ε_o and μ_o . The phase velocity is,

 $v = \frac{1}{(\epsilon \mu)^{1/2}} = \frac{c}{(\epsilon_r \mu_r)^{1/2}} = \frac{c}{n},$ (28-38)

• *n* is index of refraction:

 $n=(\epsilon,\mu_r)^{1/2}.$

(28-39)

• In nonmagnetic media $n = \epsilon_r^{1/2}$. (28-

(28-40)

- The characteristic impedance of medium is

 $Z = \frac{E}{H} = \left(\frac{\mu}{\epsilon}\right)^{1/2} = 377 \left(\frac{\mu_r}{\epsilon_r}\right)^{1/2} \quad \text{ohms.}$ (28-41)

• The electric and magnetic energy densities are again equal $\frac{\epsilon E^2/2}{\mu H^2/2} = 1$, (28-42)

Time-averaged enery density is

$$\mathcal{E}'_{av} = \frac{\epsilon E_{rms}^2}{2} + \frac{\mu H_{rms}^2}{2} = \epsilon E_{rms}^2 = \mu H_{rms}^2.$$
 (28-43)

The poyinting vector **ExH** points again in the direction of $\mathscr{G}_{av} = \frac{1}{2} \operatorname{Re} (EH^*) \hat{z} = \left(\frac{\epsilon}{\mu}\right)^{1/2} E_{rms}^2 \hat{z}$ (28-44)

$$\mathscr{S}_{av} \approx \frac{(\epsilon_r/\mu_r)^{1/2} E_{rms}^2}{377} \hat{z} \quad watts/meter^2$$
 (28-45)

$$= \frac{1}{\epsilon \mu} \epsilon E_{\rm rms}^2 \hat{z} = v \epsilon E_{\rm rms}^2 \hat{z}.$$
(28-46)

The time-aveaged Poyinting vector is equal to total energy density multiplied by the phase velocity

Uniform Plane Electromagnetic Waves in Conductor- Complex Wave Number ($k = \beta - j\alpha$)

- In the conductin medium $k^2 = \frac{\epsilon_r \mu_r}{k_r^2} \left(1 j \frac{\sigma}{\omega \epsilon}\right),$ (28-52)
- So k is complex. Is is the costum to set

 $k = \beta - j\alpha$ and then $E = E_m \exp(-\alpha z) \exp j(\omega t - \beta z)$, (28-53)

• $1/\alpha$ is the attenuation distance or the skin dept δ over which the attenuation decreases by a facter of *e*. So,

 $\alpha = \frac{1}{\delta}, \qquad (28-54)$ $\beta = \frac{1}{\delta} = \frac{2\pi}{\delta}, \qquad (28-55)$

And phase velocity is

 $v = \frac{\omega}{\beta}$. (28-56)

• Let us find α and β in terms of ϵ_r , μ_r and σ . First we set:

 $\mathscr{D} = \frac{\sigma}{\omega \epsilon} = \left| \frac{\sigma E}{\epsilon \, \partial E / \partial t} \right| = \left| \frac{\sigma E}{\partial D / \partial t} \right| \approx 377 \, \frac{\sigma \tilde{\lambda}_0}{\epsilon_r}.$ (28-57)

 $\mathcal{D} = \tan l, \tag{28-58}$

This is called loss angle of the medium

Thus

$$k^{2} = (\beta - j\alpha)^{2} = \left(\frac{\epsilon_{r}\mu_{r}}{\lambda_{0}^{2}}\right)(1 - j\mathcal{D}), \quad (28-60)$$

$$\alpha = \frac{1}{\lambda_{0}} \left(\frac{\epsilon_{r}\mu_{r}}{2}\right)^{1/2} [(1 + \mathcal{D}^{2})^{1/2} - 1]^{1/2}, \quad (28-61)$$

$$\beta = \frac{1}{\lambda_{0}} \left(\frac{\epsilon_{r}\mu_{r}}{2}\right)^{1/2} [(1 + \mathcal{D}^{2})^{1/2} + 1]^{1/2}, \quad (28-62)$$

$$k = \frac{(\epsilon_{r}\mu_{r})^{1/2}}{\lambda_{0}} (1 + \mathcal{D}^{2})^{1/4} \exp\left(-j \arctan\frac{\alpha}{\beta}\right). \quad (28-63)$$
For D<<1 (low loss dielectric)

$$\alpha \approx \frac{(\epsilon_{r}\mu_{r})^{1/2}\mathcal{D}}{2\lambda_{0}} = \left(\frac{\mu_{r}}{\epsilon_{r}}\right)^{1/2} \frac{\sigma c \mu_{0}}{2}, \quad (28-64)$$

$$\beta \approx \frac{(\epsilon_{r}\mu_{r})^{1/2}}{\lambda_{0}}, \quad v = \frac{\omega}{\beta} \approx \frac{c}{(\epsilon_{r}\mu_{r})^{1/2}}. \quad (28-65)$$
For D>>1 (good conductor)

$$k^{2} = -i\mathcal{D} \frac{\epsilon_{r}\mu_{r}}{\epsilon_{r}} = -i\frac{\sigma}{\epsilon_{r}} \epsilon_{r}\mu_{r}\mathcal{D}^{2} \epsilon_{r}\mu_{r} = -i\sigma\mu\omega, \quad (28-66), \quad \mu = \frac{c}{\epsilon_{r}} = \frac{c\beta}{2}$$

$$e^{2} = -j \mathscr{D} \frac{\epsilon_{r} \mu_{r}}{\lambda_{0}^{2}} = -j \frac{\sigma}{\omega \epsilon} \epsilon_{r} \mu_{r} \omega^{2} \epsilon_{0} \mu_{0} = -j \sigma \mu \omega, \qquad (28-66) \quad n = \frac{c}{\omega/\beta} = \frac{c\beta}{\omega} = c \left(\frac{\sigma \mu}{2\omega}\right)^{1/2}$$

$$k = \left(\frac{\sigma \mu \omega}{2}\right)^{1/2} (1-j), \qquad (28-67)$$

$$\alpha = \beta = \left(\frac{\sigma \mu \omega}{2}\right)^{1/2}. \qquad (28-68)$$

Uniform Plane Electromagnetic Waves in Conductor- Characteristic Impedance

 The characteristic impedance of a conducting medium is complex:

$$Z = \frac{E}{H} = \frac{k}{\omega\epsilon - j\sigma} = \frac{\omega\mu}{k}$$
(28-70)
$$= \left(\frac{\mu}{\epsilon}\right)^{1/2} \frac{\exp j \arctan\left(\frac{\alpha}{\beta}\right)}{\left(1 + \mathcal{D}^2\right)^{1/4}} \approx 377 \left(\frac{\mu_r}{\epsilon_r}\right)^{1/2} \frac{\exp j \arctan\left(\frac{\alpha}{\beta}\right)}{\left(1 + \mathcal{D}^2\right)^{1/4}} \text{ ohms}$$
(28-71)

- This menans that E and H are not in phase:
 - $\frac{E}{H} = \frac{\omega\mu}{\beta j\alpha}, \qquad (28-72) \quad \theta = \arctan\frac{\alpha}{\beta}. \qquad (28-73)$
- Therefore

$$\boldsymbol{E} = \boldsymbol{E}_m \exp\left(-\alpha z\right) \exp j(\omega t - \beta z), \qquad (28-74)$$

$$\boldsymbol{H} = \boldsymbol{H}_m \exp\left(-\alpha z\right) \exp j(\omega t - \beta z - \theta), \qquad (28-75)$$

With

$$\frac{E_m}{H_m} = \frac{\omega\mu}{k} = \left(\frac{\mu}{\epsilon}\right)^{1/2} \frac{1}{(1+\mathscr{D}^2)^{1/4}} = \left(\frac{\mu_0}{\epsilon_0}\right)^{1/2} \left(\frac{\mu_r}{\epsilon_r}\right)^{1/2} \frac{1}{(1+\mathscr{D}^2)^{1/4}} \quad (28-76)$$
$$\approx 377 \left(\frac{\mu_r}{\epsilon_r}\right)^{1/2} \frac{1}{(1+\mathscr{D}^2)^{1/4}} \quad \text{ohms.} \qquad (28-77)$$

Uniform Plane Electromagnetic Waves in Conductor- Energy Density

 Time-averaged electric and magnetic energy densities are in ratio

 $\frac{\mathscr{C}'_{e}}{\mathscr{C}'_{m}} = \frac{\epsilon E_{\rm rms}^{2}/2}{\mu H_{\rm rms}^{2}/2} = \frac{1}{(1+\mathscr{D}^{2})^{1/2}}.$ (28-78)

- There is less electric energy than magnetic energy because the conductivity both decreases E and adds conduction current to the displacement current, which increase H.
- The time-averaged total energy density is;

 $\frac{1}{2}(\epsilon E_{\rm rms}^2 + \mu H_{\rm rms}^2) \exp(-2\alpha z) = \frac{1}{2}(\epsilon E_{\rm rms}^2)[1 + (1 + \mathcal{D}^2)^{1/2}] \exp(-2\alpha z).$ (28-79)

Poyinting Theorem

- We referred to the Poyinting vector $\mathcal{G} = \mathbf{E} \times \mathbf{H}$ (28-80)
- This vector is of great theoretical and practical interest. To prove Poyinting Theorem, first we use vector identity

 $\nabla \cdot (E \times H) = H \cdot (\nabla \times E) - E \cdot (\nabla \times H).$ (28-81)

- In HILS medium $\nabla \cdot (E \times H) = -H \cdot \mu \frac{\partial H}{\partial t} - E \cdot \left(\epsilon \frac{\partial E}{\partial t} + J_f\right)$ (28-82) $= -\frac{\partial}{\partial t} \left(\frac{\epsilon E^2}{2} + \frac{\mu H^2}{2}\right) - E \cdot J_f.$ (28-83)
- Now we change the sign and integrate over a volume v of a finite extent and of a surface of area A, and finally apply the divergence theorem on the left then;

$$-\int_{\mathscr{A}} (\boldsymbol{E} \times \boldsymbol{H}) \cdot d\boldsymbol{\mathscr{A}} = \frac{d}{dt} \int_{\boldsymbol{v}} \left(\frac{\boldsymbol{\epsilon} \boldsymbol{E}^2}{2} + \frac{\mu \boldsymbol{H}^2}{2} \right) d\boldsymbol{v} + \int_{\boldsymbol{v}} \boldsymbol{E} \cdot \boldsymbol{J}_f \, d\boldsymbol{v}. \quad (28-84)$$

is the total power flowing out of closed surface of area A

The time-averaged magnetute of the poyinting vector is

$$\mathcal{S}_{av} = \frac{1}{2} \operatorname{Re} \left\{ \left[E_m \exp\left(-\alpha z\right) \exp j(\omega t - \beta z) \right. \\ \left. \times H_m \exp\left(-\alpha z\right) \exp j(-\omega t + \beta z + \theta) \right] \right\}$$
(28-86)

 $= \frac{1}{2} E_m H_m \cos \theta \exp \left(-2\alpha z\right), \tag{28-87}$

Where θ is defined as

 $\cos \theta = \frac{\beta}{(\alpha^2 + \beta^2)^{1/2}}.$ (28-88)

 $\mathscr{S}_{av} = \left(\frac{\epsilon}{\mu}\right)^{1/2} (1 + \mathscr{D}^2)^{1/4} E_{rms}^2 \cos\theta \exp\left(-2\alpha z\right)$ (28-89) $\approx \frac{1}{377} \left(\frac{\epsilon_r}{\mu_r}\right)^{1/2} (1 + \mathscr{D}^2)^{1/4} E_{rms}^2 \cos\theta \exp\left(-2\alpha z\right).$ (28-90)

 $\mathcal{S}_{av} = (\text{time-averaged energy density}) \times (\text{phase velocity})$ (28-91)

It is casy to show that